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Modeling quasi-dark states with temporal coupled-mode theory

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Abstract: Coupled resonators are commonly used to achieve tailored spectral responses and allow novel functionalities in a broad range of applications. The Temporal Coupled-Mode Theory (TCMT) provides a simple and general tool that is widely used to model these devices. Relying on TCMT to model coupled resonators might however be misleading in some circumstances due to the lumped-element nature of the model. In this article, we report an important limitation of TCMT related to the prediction of dark states. Studying a coupled system composed of three microring resonators, we demonstrate that TCMT predicts the existence of a dark state that is in disagreement with experimental observations and with the more general results obtained with the Transfer Matrix Method (TMM) and the Finite-Difference Time-Domain (FDTD) simulations. We identify the limitation in the TCMT model to be related to the mechanism of excitation/decay of the supermodes and we propose a correction that effectively reconciles the model with expected results. Our discussion based on coupled microring resonators can be useful for other electromagnetic resonant systems due to the generality and far-reach of the TCMT formalism.

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1. Introduction

Resonant structures are ubiquitous in nanophotonics [1–8] and their wide success is facilitated by the use of powerful and simple mathematical tools such as the transfer matrix method (TMM) [9] and temporal coupled-mode theory (TCMT) [10–12]. In TMM, the response of a resonant system is calculated directly from the combined interference of light propagating through multiple optical paths and therefore it is suitable to describe systems in which these optical paths are well known, such as optical resonators composed of waveguides or using free-space optics. In TCMT, on the other hand, the supermodes of a resonant system are calculated from the perturbative coupling of lumped resonators and, in addition to conventional resonators in waveguides and free-space, it appeals to a variety of less conventional electromagnetic resonant structures [13–16]. As a perturbative model TCMT is however limited to low-loss weakly coupled systems excited around their resonant frequency (small detuning), while TMM covers a broader range of parameters.

When both TMM and TCMT models are possible, the latter often provides a simpler formulation. For instance, TMM calculations can be cumbersome for resonant structures that allow coupling between counter-propagating modes [5, 17, 18], while TCMT can provide a more straightforward approach [4, 19]. Furthermore the TCMT formalism, based on time differential equations, provides a very simple tool to evaluate dynamic responses of resonators [1, 2]. A time-dependent TMM model is also possible [20] but at the expense of complex calculations when dealing with coupled resonators, which have been increasingly deployed for optical modu-

lation [1–3], optical computing [21,22] and in the study of dynamic phenomena such as nonlinear oscillations in silicon waveguides [7, 8].

An important application of coupled resonators is the generation and control of optical dark states. A resonant state is "dark" when it cannot be excited due to the completely destructive interference of light in the optical path connecting the resonator to external light channels [23, 24]. A slight imbalance in such destructive interference can lead to a weak effective coupling between the resonator and the external channels, originating a high quality factor (high-Q) resonance instead, or quasi-dark state. The transition between dark and quasi-dark states have been investigated for several applications including light storage [25], lasers [26], optical modulation [27] and wireless energy transfer [28], and TCMT has been often used as the modeling tool [26–28].

Here we show that TCMT might however fail to describe coupled resonators presenting quasi-dark states. We investigate a coupled device composed of three ring resonators and we show that the TCMT model predicts a dark state that is in contrast with experimental observations, TMM and FDTD simulations, which yield a high-Q quasi-dark state instead. We derive a modified TCMT model that allows the proper excitation of the quasi-dark state and reconciles the different methods.

2. Three-ring coupled resonator

The experimental realization of the three-ring coupled resonator and its spectral response are presented in Fig. 1(a-d). The device consists of two identical microring resonators coupled to a third dissimilar microring that is coupled to a bus waveguide (Fig. 1(a)). The transmission spectrum in Fig. 1(b) shows a triplet resulting from the coupling-induced mode-splitting when the three rings are degenerate, while the light distribution in each of the three resonances is illustrated in the infra-red (IR) micrographs of Fig. 1(c). The lateral resonances of the triplet have similar Q-factor (14,000) and extinction ratio, whereas the central resonance constitutes a so-called quasi-dark state with significantly higher Q-factor (66,000) as the light is mostly localized in the embedded rings, effectively reducing the extrinsic (coupling) losses. A typical anti-crossing diagram (Fig. 1(d)) is obtained from the transmission spectrum when the detuning between outer and embedded rings is controlled by means of an integrated microheater (H_1 in the inset micrograph of Fig. 1(d)). The experimental anti-crossing is subjected to an overall red-shift of the resonances due to thermal crosstalk, but the evolution of the supermodes remain clear. The device was fabricated in a standard silicon-on-insulator (SOI) platform with typical dimensions used for silicon channel waveguides [29]: 450-nm by 220-nm waveguides for quasi-TE mode operation with microring radii $R_1 = 20 \mu\text{m}$, $R_2 = R_3 = 5 \mu\text{m}$, and 200-nm coupling gap between outer and embedded microrings and between the outer microring and the bus waveguide. The microheaters and contact pads consist of 100-nm Ni-Cr and 2/200-nm Ti/Au films fabricated in a post-process step and the measured electrical resistance of microheater H_1 is 130 ohm.

The spectral features obtained experimentally are well reproduced by the TMM model (Fig. 1(e-h)) — see Appendix A for the TMM equations and for numerical values of the parameters used to obtain the plots in Fig. 1. The transmission spectrum (Fig. 1(f)) shows a similar triplet with resonance-splitting dictated by the coupling between embedded and outer ring resonators (k_2) and with a high-Q central resonance. The intracavity power for each ring (Fig. 1(g)) — normalized to s_{in} — provides a quantitative assessment of the power distributions observed in the IR-micrographs: for both lateral resonances light circulates in all the three rings, while for the central resonance light is localized within the embedded rings resulting in a small effective coupling to the bus waveguide and high power enhancement. Finally, the TMM anti-crossing diagram of Fig. 1(h) provides the expected evolution of the supermodes in the absence of thermal crosstalk.

Unlike TMM, however, the TCMT model disagrees with the experimental observations as it

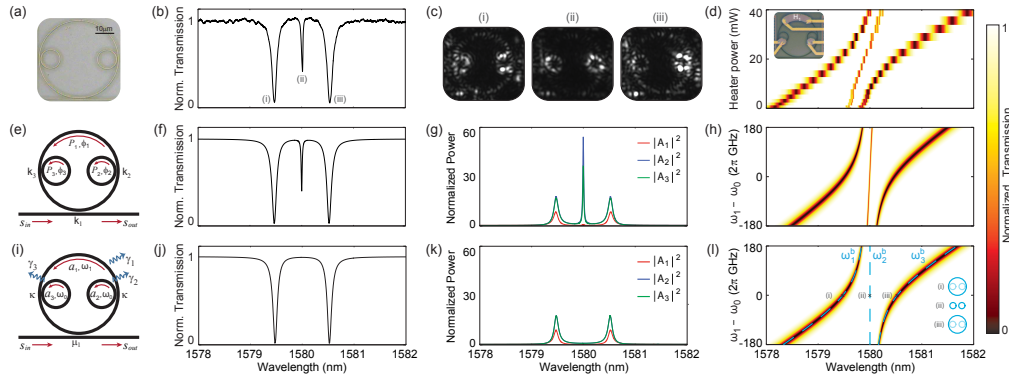


Fig. 1. Experimental data (a-d), TMM model (e-h) and TCMT model (i-l) of a three-ring resonator system. (a) Fabricated device and (b) transmission spectrum showing a triplet with high-Q quasi-dark state in the center when the three rings are degenerate. (c) IR-micrograph of the scattered light at each resonance. (d) Anti-crossing obtained when the outer ring is detuned using a microheater (H_1 in the inset micrograph). An overall red-shift is present due to thermal crosstalk affecting the embedded rings. (e) TMM parameters: s_{in}/s_{out} are the input/output fields; k_1 and k_2 are coupling coefficients and ϕ_i and P_i are the accumulated phase and attenuation in each microring, respectively. (f) TMM triplet similar to the experimental observation and (g) intracavity power spectrum with high power enhancement for the central resonance, in which case light is confined to the embedded rings. $|A_1|^2$, $|A_2|^2$ and $|A_3|^2$ represent the power circulating in the outer ring, first and second embedded rings, respectively. The blue and green curves closely overlap. (h) TMM anti-crossing showing the evolution of the supermode resonances in the absence of thermal crosstalk. (i) TCMT model (parameters described in the text). In contrast to the experimental data and TMM, no central resonance is observed in the (j) transmission spectrum and (k) intracavity power spectrum. (l) TCMT anti-crossing obtained from the transmission spectrum and from the eigenvalues of Ω (dashed-blue lines). The central mode is predicted by the eigenvalues but not excited, constituting a dark state in the TCMT model. Inset: spatial distribution of each supermode at degeneracy, representing the eigenvectors of Ω .

predicts a dark state for the central supermode (Fig. 1(i-l)). No central resonance notch appears in the transmission spectrum (Fig. 1(j)) and no light circulates in the cavity (Fig. 1(k)) since the supermode cannot be excited by the incoming light. These results are calculated using a general formulation of the orthogonal TCMT [11] (see Appendix B),

$$\frac{d\vec{a}}{dt} = (j\Omega - \Gamma) \cdot \vec{a} + K^T \cdot \vec{s}_{in} \quad (1)$$

$$\vec{s}_{out} = C \cdot \vec{s}_{in} + K \cdot \vec{a} \quad (2)$$

with the following parameters for the three-ring model (Fig. 1(i)) [27, 28]: the mode amplitude of the individual resonators are grouped in the mode vector $\vec{a} = (a_1 \ a_2 \ a_3)^T$ and their bare resonance frequencies (ω_1 for the external ring and ω_0 for the identical embedded rings) and mutual coupling (κ) constitute the system matrix

$$\Omega = \begin{pmatrix} \omega_1 & \kappa & \kappa \\ \kappa & \omega_0 & 0 \\ \kappa & 0 & \omega_0 \end{pmatrix}; \quad (3)$$

the single bus waveguide is described by the incoming and outgoing power amplitudes s_{in} and s_{out} and requires $C = 1$; the coupling between the bus waveguide and the resonant system, occurring

only through the outer ring, is represented by the coupling vector

$$K = \begin{pmatrix} j\mu_1 & 0 & 0 \end{pmatrix} \quad (4)$$

where μ_1 represents the coupling of the first resonator to the bus waveguide; the decay matrix $\Gamma = \Gamma^{loss} + \Gamma^{port}$ completes the model, accounting for the intrinsic losses in each resonator, $\Gamma^{loss} = \text{diag}(\gamma_1 \ \gamma_2 \ \gamma_3)$ and for the extrinsic loss term $\Gamma^{port} = \text{diag}(\mu_1^2/2 \ 0 \ 0)$ according to eq.(B5). The transmission spectrum is calculated as $|s_{out}/s_{in}|^2$ while the intracavity power spectra are calculated using eq.(B3).

The prediction of a dark state by TCMT can be understood considering the interaction between the supermodes of the coupled system and the bus waveguide. First, we calculate the eigenvalues and normalized eigenvectors of Ω , which give the supermodes' resonance frequencies ω_i^b 's and mode amplitudes b_i 's. The ω_i^b 's are depicted in Fig. 1(l) (blue traces) and they follow closely the spectral evolution expected from the experimental results and TMM. At degeneracy, the eigenfrequencies and eigenmodes are

$$\begin{aligned} \omega_1^b &= \omega_0 + \sqrt{2} \ \kappa \quad , \quad b_1 = \begin{pmatrix} -\frac{1}{\sqrt{2}} & \frac{1}{2} & \frac{1}{2} \end{pmatrix}^T \\ \omega_2^b &= \omega_0 \quad , \quad b_2 = \begin{pmatrix} 0 & -\frac{1}{\sqrt{2}} & \frac{1}{\sqrt{2}} \end{pmatrix}^T \\ \omega_3^b &= \omega_0 - \sqrt{2} \ \kappa \quad , \quad b_3 = \begin{pmatrix} \frac{1}{\sqrt{2}} & \frac{1}{2} & \frac{1}{2} \end{pmatrix}^T . \end{aligned} \quad (5)$$

These expressions show that b_2 , the supermode corresponding to the (quasi-)dark state, is completely confined to the embedded rings while b_1 and b_3 have components in the outer ring as illustrated in the inset of Fig. 1(l). Since b_2 vanishes in the outer ring it cannot be excited by the incoming light which only couples to a_1 (see eq.(4)). The effective zero drive for supermode b_2 can be directly seen rewriting K in the coupled basis,

$$K^b = j \frac{\mu_1}{\sqrt{2}} \begin{pmatrix} -1 & 0 & 1 \end{pmatrix} \quad (6)$$

calculated as $K^b = K \cdot (S^{-1})^T$, where S is the similarity matrix formed by the column eigenvectors of eq.(5). K^b represents the coupling between the input/output power amplitudes (s_{in}/s_{out}) and supermodes b_1 , b_2 and b_3 and it shows that the coupling to b_2 is effectively zero.

FDTD simulations of two distinct three-ring designs allow to understand the limitations of the TCMT model and how it can be modified to properly describe the quasi-dark state. The transmission spectrum and the mode profiles of each supermode are presented in Fig. 2(a,c) for a design similar to the one described in Fig. 1 with two embedded rings coupled to the outer ring at different positions, while Fig. 2(b,d) shows these results for a design where both rings are coupled to the outer ring at the same point. The FDTD simulation for the first design is consistent with the previously discussed experimental and TMM results as it also predicts the excitation of the quasi-dark state. In the second design, on the other hand, the supermode is not excited and constitutes therefore an effective dark state in agreement with the TCMT prediction. In a lumped element model such as TCMT, however, these two designs are equivalent: two identical resonators weakly coupled to a third one which in turn is coupled to a bus waveguide, with system matrix and coupling vector given by eq.(3,4) and supermodes given by eq.(5).

The fundamental difference between the two designs which not accounted in the TCMT model lies on the effect of the embedded rings in the roundtrip phase of the outer ring. In the absence of embedded rings the outer ring is resonant at ω_0 and its accumulated roundtrip phase is a multiple of 2π . When the two embedded rings couple to the outer ring at different positions, each of them introduces a zero or π phase-shift depending on its coupling regime (undercoupled or overcoupled, respectively [30]) and the accumulated roundtrip phase in the outer ring remains a multiple of 2π . Even though no resonant light circulates in the outer ring as indicated by b_2

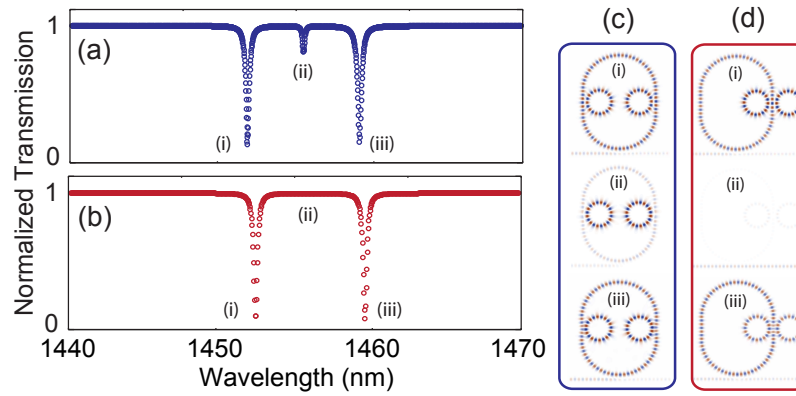


Fig. 2. 2D-FDTD simulations. (a,b) Transmission spectrum and (c,d) steady-state electric field amplitude of the supermodes of the three-ring device at degeneracy in two different configurations. (a,c) When the embedded rings coupled to the outer ring at different positions a weak field circulates in the outer ring allowing the excitation of the quasi-dark state (ii). (b,d) When the embedded rings couple to the outer ring at the same position the destructive interference in the outer ring prevents the excitation of supermode (ii), originating a dark state.

in eq.(5) an amount of non-resonant light is able to propagate over the outer ring to feed the embedded rings as illustrated in Fig. 2(c-ii). In other words, the outer ring acts as a waveguide in this situation, allowing the communication between the supermode b_2 confined in the embedded rings and the bus waveguide. The contribution of this non-resonant light is not considered in the TCMT model, which only accounts for the resonant mode amplitudes. On the other hand, when the coupling between rings occurs in the exact same point a π phase shift is introduced in the accumulated roundtrip phase of the outer ring, resulting in destructive interference and preventing the excitation of b_2 , as shown in Fig. 2(d-ii). Detailed information regarding the parameters used in the FDTD simulations can be found in Appendix C. Notice that we used a racetrack as the outer resonator to assure the coupling between outer and embedded rings is the same in both designs.

The TCMT model can nonetheless be altered to deliver a description of the three-ring design that allows for the excitation of the quasi-dark state b_2 . This is accomplished with a modified coupling vector which will be derived in the next section.

3. TCMT with modified coupling vector

In this section, we propose a modified TCMT model that incorporates new terms to the coupling vector K and yields results in agreement with TMM and FDTD. The coupling vector for the three-ring system can be written in its most general form, according to eqs.(B4, B5), as $K = j(\mu_1 \ \mu_2 \ \mu_3)$ where $\mu_i \in \mathbb{R}$. A modification of K requires a modification of Γ^{port} , whose components are given by $\Gamma_{ij}^{port} = \frac{\mu_i \mu_j}{2}$. Therefore, modifying the TCMT model reduces to deducing the correct expressions for the coupling terms μ_i .

We seek to write the μ_i 's in terms of the power coupling coefficients k_i 's defined in Fig. 1(e) in order to establish a direct correspondence between TCMT and the power coupling parameters used in TMM. The first term μ_1 is the usual power-energy coupling coefficient for a bus-ring configuration [12], written in terms of k_1 as

$$\mu_1 = k_1 \sqrt{\frac{v_g}{L_1}} \quad (7)$$

where v_g is the group velocity in the outer ring (we will assume the same group velocity for all rings). In the coupled basis, the general coupling vector is

$$K^b \equiv \begin{pmatrix} K_1^b \\ K_2^b \\ K_3^b \end{pmatrix}^T = \begin{pmatrix} -j\frac{1}{2}(\sqrt{2}\mu_1 - \mu_2 - \mu_3) \\ -j\frac{(\mu_2 - \mu_3)}{\sqrt{2}} \\ j\frac{1}{2}(\sqrt{2}\mu_1 + \mu_2 + \mu_3) \end{pmatrix}^T \quad (8)$$

It is expected that supermodes b_1 and b_3 be equally coupled to the bus waveguide ($|K_1^b|^2 = |K_3^b|^2$) as they have the same mode profile, thus requiring μ_2 and μ_3 to satisfy $\mu_3 = -\mu_2$. This allows to simplify K^b to

$$K^b = j\frac{1}{\sqrt{2}} \begin{pmatrix} -\mu_1 & -2\mu_2 & \mu_1 \end{pmatrix}. \quad (9)$$

This expression gives the same coupling to b_1 and b_3 as eq.(6) which was already in agreement with the expected results. As for mode b_2 , it can now be excited by a non-null μ_2 .

The term μ_2 represents the indirect coupling between bus waveguide and embedded rings and its dependence with k_1 and k_2 is determined using power conservation [11, 12]. Consider a lossless system ($\Gamma^{\text{loss}} = 0$) with mode b_2 excited to energy $|b_2^0|^2$ at $t = 0$. With no incoming light ($s_{in} = 0$) the energy in the resonator decays and the power flowing through the output port is $|s_{out}(t)|^2 = 2\mu_2^2 |b_2(t)|^2$. The same scenario can be described using a power-normalized amplitude $B_2(t)$ that couples to the outgoing wave $s_{out}(t)$ through a power coupling coefficient defined as k_b so that $|s_{out}(t)|^2 = k_b^2 |B_2(t)|^2$. The equivalence between the two pictures requires

$$2\mu_2^2 |b_2(t)|^2 = k_b^2 |B_2(t)|^2. \quad (10)$$

On the other hand, the relation between circulating power and stored energy given by eq.(B3) requires

$$|B_2(t)|^2 = |b_2(t)|^2 \frac{v_g}{2L_2} \quad (11)$$

where L_2 is the length of each identical embedded ring so that $2L_2$ is the effective length of supermode b_2 . Eqs.(10, 11) allow to write μ_2 in terms of k_b :

$$\mu_2 = \frac{k_b}{2} \sqrt{\frac{v_g}{L_2}} \quad (12)$$

Finally, the power coupling coefficient k_b is given by $\frac{k_1 k_2}{2}$, as follows. The supermode confined to the embedded rings, with power $|B_2|^2$, is fed by a certain amount of power circulating in the outer ring $|A_1^\pi|^2$ by means of k_2 , so that

$$|B_2|^2 = k_2^2 |A_1^\pi|^2. \quad (13)$$

At frequency ω_2^b , for which this correction is derived, $|A_1^\pi|^2$ is the circulating power in a microring out-of-resonance and can be estimated using TMM along with the TCMT assumptions of low loss ($P_1 \rightarrow 1$) and weak coupling ($k_1 \ll 1$):

$$|A_1|^2 = \left| \frac{jk_1 P_1 e^{j\phi_1}}{1 - t_1 P_1 e^{j\phi_1}} \right|^2 |s_{in}|^2 \xrightarrow[k_1 \ll 1, P_1 \rightarrow 1]{\phi_1 \rightarrow \pi} |A_1^\pi|^2 = \frac{k_1^2}{4} |s_{in}|^2 \quad (14)$$

Combining eq.(13) and eq.(14) and reminding that k_b was defined as the coupling coefficient between B_2 and the bus waveguide, we have $k_b = \frac{k_1 k_2}{2}$.

The modification of the TCMT model is therefore complete, consisting of a new coupling vector $K = j(\mu_1 \mu_2 - \mu_2)$ and additional elements $\Gamma_{ij}^{\text{port}} = \frac{\mu_i \mu_j}{2}$ to the decay matrix, where

$$\mu_1 = k_1 \sqrt{\frac{v_g}{L_1}}, \quad \mu_2 = \frac{k_1 k_2}{4} \sqrt{\frac{v_g}{L_2}}. \quad (15)$$

The correct response can be obtained by using these parameters in eq.(1,2).

4. Validation of the modified-TCMT model

We validate the modified-TCMT (m-TCMT) steady-state solution by its comparison with TMM. The transmission spectrum of (Fig. 3(a)) shows that m-TCMT closely reproduces the experimental and TMM traces, including a clear high-Q central resonance associated with an excited quasi-dark state. Although the intracavity power spectra calculated with m-TCMT (Fig. 3(b)) and with TMM (Fig. 3(c)) reveal some differences for the power circulating in the embedded rings (inset figures), their average power for supermode b_2 , calculated as $(|A_2|^2 + |A_3|^2)/2$, are in very good agreement (Fig. 3(d)). The asymmetry between blue and green traces in m-TCMT (inset of Fig. 3(b)) reflects the asymmetry of the cross-decay terms Γ_{12}^{port} and Γ_{13}^{port} in

$$\Gamma^{port} = \begin{pmatrix} \frac{\mu_1^2}{2} & \frac{\mu_1 \mu_2}{2} & -\frac{\mu_1 \mu_2}{2} \\ \frac{\mu_1 \mu_2}{2} & \frac{\mu_2^2}{2} & -\frac{\mu_2^2}{2} \\ -\frac{\mu_1 \mu_2}{2} & -\frac{\mu_2^2}{2} & \frac{\mu_3^2}{2} \end{pmatrix} \quad (16)$$

due to the fact that $\mu_3 = -\mu_2$ in our model. On the other hand, the power imbalance between $|A_2|^2$ and $|A_3|^2$ in TMM (inset of Fig. 3(c)) reflects the fact that light arrives at the second embedded ring modified by the resonance of the first one, as can be seen in the expression

$$\frac{|A_3|^2}{|A_2|^2} = P_1 |\chi_2|^2 \quad (17)$$

where P_1 is the roundtrip attenuation factor of the outer ring and χ_2 is the complex transmission of the first embedded ring (see Appendix A). The new terms in the m-TCMT equations depend on the coupling coefficients k_1 and k_2 , therefore the model must be validated over a wide range of

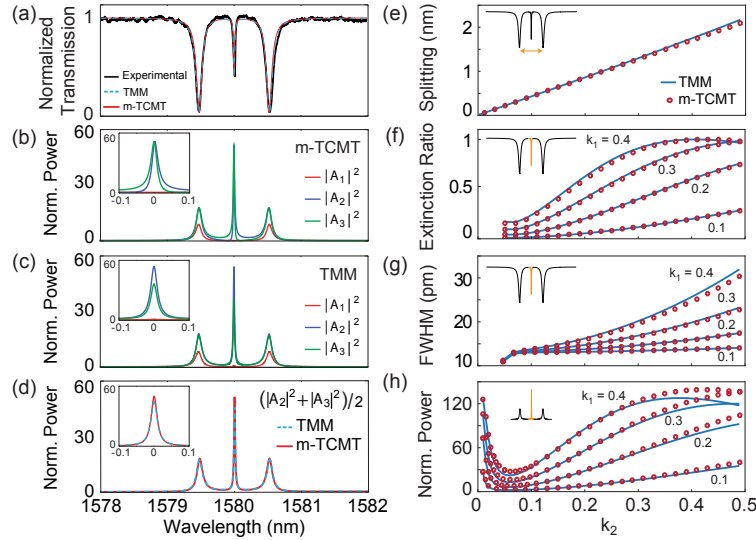


Fig. 3. Comparison between modified-TCMT (m-TCMT) and TMM. (a) Transmission spectrum and (b,c) intracavity power spectrum calculated with the parameters used in Fig. 1. Insets: detail of the central peaks. (d) Average intracavity power for the quasi-dark state calculated with TMM and m-TCMT. (e) Resonance splitting for the triplet and (f) extinction ratio, (g) linewidth and (h) average intracavity power for the quasi-dark state calculated for various coupling coefficients. The m-TCMT calculations agree with TMM for a wide range of coupling strengths, while in the standard TCMT curves (f-h) would vanish.

these parameters. The m-TCMT and TMM models yield very close predictions for various values of coupling coefficients as demonstrated in Fig. 3(e-h), which shows results for the resonance splitting (Fig. 3(e)), the extinction ratio and linewidth of the quasi-dark state resonance (Fig. 3(f,g)) and its average intracavity power (Fig. 3(h)). Particularly, the agreement between TMM and m-TCMT for the average intracavity power indicates that the power imbalance captured in the TMM model do not significantly affect the total power in the supermode. The two models show slight discrepancies only for combinations of large coupling strengths, when the weak-coupling assumption of TCMT starts to fail. Notice that, except for the resonance splitting, the calculated quantities would vanish in the standard TCMT model. A similar comparison for the lateral resonances of the triplet (supermodes b_1 and b_3) is unnecessary as they were already well described by the standard TCMT and their steady-state values are not affected by μ_2 , as predicted by eq.(9). The results presented in Fig. 3(e-h) were calculated assuming the same ring radii and effective index used in previous plots and presented in Appendix A, whilst the attenuation parameters where $P_1 = 0.992$ and $P_1 = 0.998$ (equivalent to $\alpha_1 = \alpha_2 = 5$ dB/cm).

In addition to the steady-state response, the m-TCMT model provides a good description of the transient behavior of the coupled system as confirmed by a comparison with 2D-FDTD simulations (Fig. 4) — see Appendix C for simulation parameters. The transient evolution of the lateral resonances (Fig. 4(b)) presents fast oscillations determined by the resonance splitting ($2\sqrt{2}\kappa$) which are also reproduced by m-TCMT (Fig. 4(c)) as a result of the additional non-diagonal terms in eq.(16). For the quasi-dark state, the FDTD transient presents a power imbalance between embedded rings similar to that observed in the TMM calculations (Fig. 4(d)).

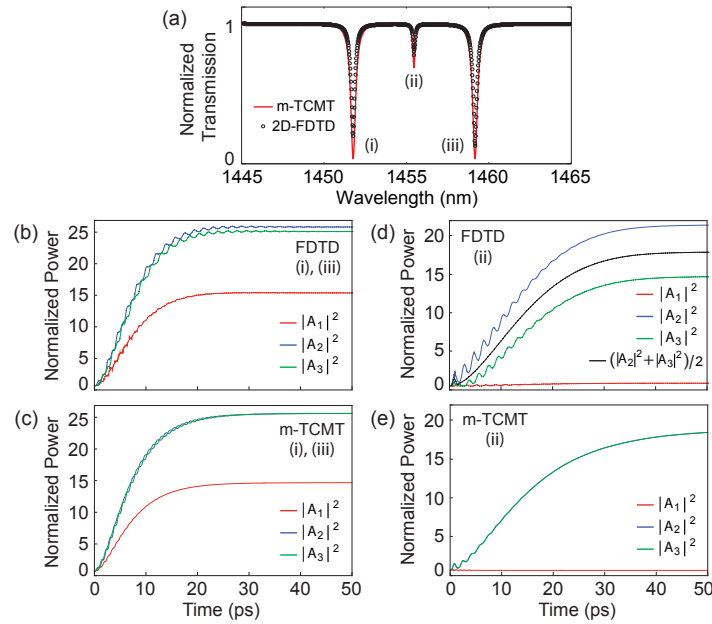


Fig. 4. Comparison between m-TCMT and 2D-FDTD simulations. (a) Transmission spectrum and (b-e) transient intracavity power evolution. The FDTD results are presented in (b) for the lateral resonances and in (d) for the quasi-dark state, while the corresponding m-TCMT solutions are presented in (c) and (e). The m-TCMT model reproduces the transient evolution for the lateral modes including the fast oscillations presented in the FDTD simulation. For the quasi-dark state it describes the average power circulating in the resonator.

Once again, it is the average power in the supermode that corresponds to the m-TCMT solution (Fig. 4(e)). This good description of the transient response of the optical supermodes makes m-TCMT a suitable model to describe dynamic perturbations on the coupled-ring system such as optical modulation through refractive index perturbation.

5. Conclusion

In this article we demonstrated an important limitation of the TCMT model: the incorrect prediction of dark-states in coupled resonators instead of the actual high-Q quasi-dark states. Through the analysis of a three-ring resonator system, we showed this inaccurate prediction occurs due to the inability of TCMT to account for non-resonant light circulating in the system. The existence of such non-resonant light can be properly accounted for by introducing extra drive terms in a modified-TCMT (m-TCMT) model.

The TCMT formalism is applied to many fields and the limitation demonstrated here might be present in distinct resonant structures. We believe our discussion will be helpful to prevent the misrepresentation of quasi-dark states and will provide insight on how the standard TCMT model can be modified to allow accurate results.

Appendix A: TMM expressions for the three-ring device

In the main text we compared the transmission and the intracavity power spectra obtained with TCMT and TMM. The TCMT results can be obtained directly from the expressions presented in the main text, while the TMM expressions are presented bellow.

The transmission spectrum (T) of the coupled ring device and the intracavity power circulating

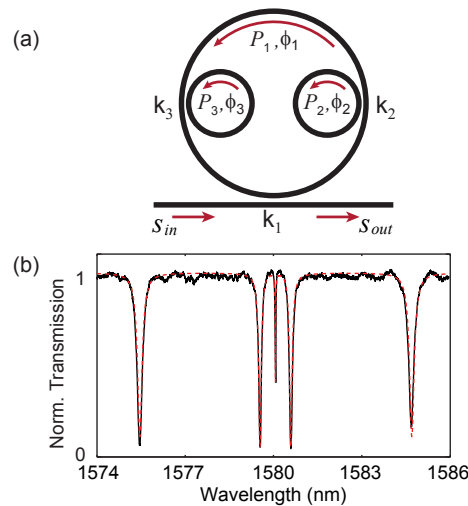


Fig. 5. (a) Schematics of the three-ring resonator with parameters used in the TMM model. (b) TMM fitting of the triplet and adjacent outer ring resonances to extract the group index for wavelengths around 1580 nm.

in each microring at steady-state (A_1, A_2 and A_3) are calculated as

$$T = \left| \frac{t_1 - P_1 e^{j\phi_1} \chi_2 \chi_3}{1 - t_1 P_1 e^{j\phi_1} \chi_2 \chi_3} \right|^2 \quad (A1)$$

$$A_1 = \left| jk_1 \frac{P_1 e^{j\phi_1} \chi_2 \chi_3}{1 - t_1 P_1 e^{j\phi_1} \chi_2 \chi_3} \right|^2 \quad (A2)$$

$$A_2 = \left| jk_1 jk_2 \frac{P_2 e^{j\phi_2}}{1 - t_2 P_2 e^{j\phi_2}} \frac{(P_1 e^{j\phi_1})^{1/4}}{1 - t_1 P_1 e^{j\phi_1} \chi_2 \chi_3} \right|^2 \quad (A3)$$

$$A_3 = \left| jk_1 jk_3 \frac{P_3 e^{j\phi_3}}{1 - t_3 P_3 e^{j\phi_3}} \frac{(P_1 e^{j\phi_1})^{3/4} \chi_2}{1 - t_1 P_1 e^{j\phi_1} \chi_2 \chi_3} \right|^2 \quad (A4)$$

where χ_2 and χ_3 account for the propagation in each embedded ring:

$$\chi_i = \frac{t_i - P_i e^{j\phi_i}}{1 - t_i P_i e^{j\phi_i}}. \quad (A5)$$

In the above expressions, $P_i = e^{\alpha_i L_i/2}$ and $\phi_i = (2\pi/\lambda)n_{eff}L_i$ are the roundtrip attenuation factor and accumulated phase, respectively. L_i and α_i are the perimeter and the distributed attenuation coefficient of microring i , n_{eff} is the effective index of the waveguide mode and λ is the free-space wavelength of the incident light. Terms t_i and k_i are the transmission and coupling coefficients and satisfy $t_i^2 + k_i^2 = 1$ (lossless coupling). In our discussion, the embedded rings are identical and therefore index $i = 3$ can be replaced by $i = 2$.

The silicon waveguides present strong dispersion that must be taken into account to provide meaningful results. We assume a first order approximation for the effective index around the wavelength of interest ($\lambda_0 = 1580$ nm):

$$n_{eff}(\lambda) \approx n_g(\lambda_0) + \lambda \partial n_{eff}(\lambda_0) / \partial \lambda \quad (A6)$$

where $n_g(\lambda_0)$ is the group index around λ_0 , calculated as 4.25 from the free spectral range of the outer ring resonances in the experimental transmission spectrum (Fig.5(b)). The differential coefficient $\partial n_{eff}(\lambda_0) / \partial \lambda$ is $-1.22 \times 10^{-3} \text{ nm}^{-1}$, yielding $n_{eff}(\lambda_0) \approx 2.30$. Bending effects in the effective index are small for the radii used (5 μm to 20 μm), therefore the same values of effective and group indices were used for all waveguides (bus waveguide, outer ring and embedded rings).

In addition to the refractive index and microring radii ($R_1 = 20 \mu\text{m}$, $R_2 = R_3 = 5 \mu\text{m}$), the coupling and attenuation parameters used to fit the experimental data are $k_1 = 0.4$, $k_2 = 0.25$, $P_1 = 0.958$ and $P_2 = 0.997$.

Appendix B: General equations of the orthogonal TCMT

A system incorporating multiple resonances and input/output ports (Fig. 6(a)) can be represented in a compact, simple manner through the TCMT formalism. The general TCMT equations describing the coupling of N resonators connected to M ports are [11]

$$\frac{d\vec{a}}{dt} = (j\Omega - \Gamma) \cdot \vec{a} + K^T \cdot \vec{s}_{in} \quad (B1)$$

$$\vec{s}_{out} = C \cdot \vec{s}_{in} + K \cdot \vec{a}. \quad (B2)$$

In the above equations, $\vec{a} = (a_1(t) \ a_2(t) \ \cdots \ a_N(t))^T$ is the mode vector comprising the N mode amplitudes a_i of the bare system, while the M ports are represented by $\vec{s}_{in} =$

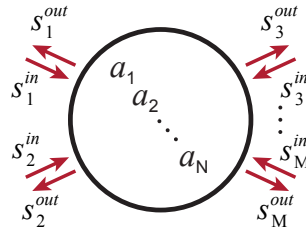


Fig. 6. Representation of a general resonant system with N resonant modes (a_i) and M input/output ports (s_i^{in}/s_i^{out}).

$(s_1^{in}(t) \ s_2^{in}(t) \ \dots \ s_M^{in}(t))^T$ and $\vec{s}_{out} = (s_1^{out}(t) \ s_2^{out}(t) \ \dots \ s_M^{out}(t))^T$, comprising the input/output power amplitudes. These amplitudes are normalized so that $|a_i(t)|^2$ represents the total energy stored in mode i at time t , while $|s_j^{in}(t)|^2$ and $|s_j^{out}(t)|^2$ represent the total power flowing *in* and *out* the system through port j . When dealing with systems of microring resonators, it is useful to consider traveling wave amplitudes $A_i(t)$ normalized such that $|A_i(t)|^2$ represents the total power flowing through any cross section of ring i . These power amplitudes are related to the mode amplitude $a_i(t)$ through [12]

$$|A_i(t)|^2 = |a_i(t)|^2 \frac{v_g^i}{L_i} \quad (B3)$$

where L_i is the length and v_g^i is the group velocity of the associated transverse mode of ring i . Matrix $\Omega_{N \times N}$ comprises the resonance frequencies ω_i of the bare modes (a_i) as its diagonal elements and the coupling between modes κ_{ij} , calculated by their overlap integral [12], as its off-diagonal elements. Matrix $\Gamma_{N \times N}$, representing the decay of the mode amplitudes, can be written as $\Gamma = \Gamma^{port} + \Gamma^{loss}$, where Γ^{port} describes the decay due to coupling to the ports and Γ^{loss} accounts for intrinsic losses. The resonant modes can be excited by the incoming light and decay into the output ports through the coupling matrix $K_{M \times N}$, while the direct power flow through ports (without the influence of resonances) is represented by the unitary and symmetric scattering matrix $C_{M \times M}$. Energy conservation and time-reversal symmetry considerations require that matrices C , K and Γ^{port} satisfy the following relations [11]:

$$CK^* = -K \quad (B4)$$

$$K^\dagger K = 2 \Gamma^{port} \quad (B5)$$

For resonant systems composed of microring resonators the loss and coupling coefficients Γ^{loss} , K and κ_{ij} can be very conveniently rewritten in terms of their equivalent TMM parameters α and k 's defined in part I [12]. First, Γ^{loss} can be written in terms of the intrinsic loss rate of each resonator γ_i as $\Gamma^{loss} = \text{diag}(\gamma_1 \ \gamma_2 \ \dots \ \gamma_N)$ and each γ_i can in turn be written in terms of the distributed loss coefficients α_i :

$$\gamma_i = \frac{v_g^i}{2} \alpha_i. \quad (B6)$$

In addition, $K = [j\mu_{qi}]$ where j is the imaginary unit and μ_{qi} is a real number representing the coupling of resonator i to port q in TCMT. Now, μ_{qi} can be expressed as

$$\mu_{qi} = k_{qi} \sqrt{\frac{v_g^i}{L_i}} \quad (B7)$$

where k_{qi} is the power coupling coefficient between port q and resonator i defined in TMM. Finally, the coupling between resonators κ_{ij} can also be written in terms of coupling coefficients k_{ij} as

$$\kappa_{ij} = k_{ij} \sqrt{\frac{v_g^i v_g^j}{L_i L_j}}. \quad (\text{B8})$$

Notice these expressions are only valid for small losses and weak coupling.

Appendix C: FDTD simulation parameters

The 2D-FDTD simulations presented in the main text were performed using the software FullWAVE [31]. In our analysis we used the TE polarization, defined such that the electric field is normal to the simulation plane. The devices were scaled down in order to reduce computation time. The bus-coupled resonator consisted of a racetrack with 3- μm bending radius and 2- μm straight sections while the embedded rings had a 1- μm bending radius. The gap between outer ring and bus waveguide and between coupled resonators were 250 nm and 200 nm, respectively, and all waveguides were 200-nm wide. The index of refraction of the surrounding background was 1.45 while different indices were used for the waveguide sections as follows: 3.48 for the bus waveguide and the racetrack; 3.497 for the embedded rings in the first design (coupling to the racetrack at different points); 3.502 for the embedded rings in the second design (coupling to the racetrack at the same point). The slight difference in refractive index between embedded rings and racetrack assures the degeneracy of all the three resonators. Notice that the index used for the embedded rings is different in each design in order to compensate for the phenomenon of coupling induced resonance frequency shift (CIFS) [32]. Losses in the system were accounted by introducing a non-null imaginary part to the refractive index of all waveguides $n_{\text{img}} = 10^{-4}$.

The transmission spectra were calculating using a pulsed simulation. A broadband pulse was launched into the bus waveguide and the outgoing light in the through port was recorded until the pulse has completely exited the simulation domain. The transmission spectra were obtained from the Fast Fourier Transform (FFT) of the time monitor in the through port. Once the spectral response was calculated, the transient response of the devices was obtained launching a continuous-wave excitation at wavelengths corresponding to each resonance of interest.

The FDTD results are compared to modified-TCMT calculations in Fig. 4 of the main text. The parameters used in the modified-TCMT calculations are: $n_g = 3.06$, $k_1 = 0.285$, $k_2 = 0.29$, $P_1 = 0.982$ and $P_2 = 0.996$.

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